High-Level Synthesis of a MPEG-4 Decoder Using SystemC

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Preface

This Master's Thesis was written at the Department of Informatics and Mathematical Modelling at the Technical University of Denmark during April 2003 - October 2003 with supervisor Hans Holten Lund, IMM and assistant supervisor Søren Forchhammer, COM.

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Abstract

This report presents a simulation model of a hardware MPEG-4 Simple@L0 compliant video decoder for behavioral synthesis. The report deals with the subset of the MPEG-4 standard that deals with low-bitrate coding of rectangular natural video. An open source project, known as XviD, serves as the basis for this project. XviD is a software implementation of a MPEG-4 compliant video CODEC (Coder and Decoder) written in C. Modeling was done using the SystemC hardware description language, and according to the rules for behavioral modeling, described in the Synopsys documentation. Synopsys synthesis tools enable behavioral synthesis and implementation. These tools were used to synthesize part of the model. A full synthesis of the model has not been performed. A major part of this project was to rewrite the XviD software CODEC to a synthesizable behavioral model in SystemC. This process involved separating the decoder from the encoder parts, replacing non-synthesizable programming structures with synthesizable alternatives, and designing an architecture that promote parallel processing. A lot of effort was put into reducing on chip memory requirements. The results of these findings will be presented with a discussion on the use of SystemC as a modeling language for behavioral synthesis. Test results will be presented with estimation of internal band width requirements. Finally some ideas are proposed on future work of the decoder.

Keywords: Behavioral modeling, SystemC, MPEG-4 decoding, behavioral synthesis, digital video.

Resume

Denne rapport præsenterer en simulerings model af en hardware MPEG-4 Simple@L0 kompatibel video dekoder. Rapporten omhandler den del af MPEG-4 standarden som har med low-bitrate kodning af rektangulær natulig video. Et open-source project, kaldet XviD, er udgangspunktet for projektet. XviD er en software implementering af en MPEG-4 video CODEC (Coder/Decoder) skrevet i C. Modellen blev implementeret i SystemC og i overenstemmelse med Synopsys dokumentationen. Synopsys syntese værktøjer muliggør høj niveau syntese og implementering. Disse værktøjer blev brugt til at syntetisere dele af modellen. Syntese af hele modellen er ikke foretaget. En stor del af projektet var omskrivning af XviD CODEC til en syntetiserbar højniveau model i SystemC. Denne process inkluderede isolering af dekoder-delen fra enkoder-delen, omskrivning af ikke-syntetiserbar kode til syntetiserbare alternativer, samt design af en arkitektur som muliggør parallelisering af beregninger. En del tid blev envidere brugt på at minimere brugen af den intern hukommelse. Resultatet af omskrivningen samt en diskussion af SystemC som modellerings sprog til højniveau syntese er inkluderet i rapporten. Test resultater er præsenteret med en estimering af interne båndbredde krav. Nogle forslag til videre arbejde med modellen er givet sidst i rapporten.

Nøgleord: Højniveau modellering, SystemC, MPEG-4 decodning, højniveau syntese, digital video.

Acronyms

2D	2-Dimensional
3D	3-Dimensional
3G	Third Generation
3GPP	Third Generation Partnership Project
ASF	Advanced Streaming Format
CBR	Constant Bit Rate
CODEC	Coder and Decoder
DCT	Discrete Cosine Transform
DVD	Digital Versatile Disc
EMS	Enhanced Message Service
FIFO	First In First Out
FPGA	Field Programmable Gate Array
GMC	Global Motion Compensation
GOV	Group of Video Object Planes
GSM	Global Systems for Mobil Communication
HDL	Hardware Description Language
IDCT	Inverse Discrete Cosine Transform
ISDN	Integrates Services Digital Network
LAN	Local Area Network
LBR	Low Bit rate Coding
MB	Macroblock
MMS	Multimedia Message Service
MP3	MPEG Audio Layer 3
MV	Motion Vector
QCIF	Quarter Common Intermediate Format
RTL	Register Transfer Level
SMS	Short Message Service
VBR	Variable Bit Rate
VCD	Video Compact Disc

VLC	Variable Length Code
VOP	Video Object Plane
WMA	Windows Media Audio
WMV	Windows Media Video V3

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1 Introduction

Short Message Service (SMS) is a service that enables GSM subscribers to send short messages, of up to 160 characters, to one another. With an estimated 15 billion SMS messages send every month, this service has proved extremely popular. The popularity of instant messaging prompted the industry to develop services beyond simple text messages. Enhanced Message Service (EMS) is an extension of the existing SMS messaging service and allows users to include pixel-animations and melodies in messages. EMS builds on existing network infrastructures and can be viewed as an intermediate stage towards full multimedia support. Multimedia Messaging Service (MMS) is the next generation of instant messaging and will support color pictures, animations, audio and video. Mobile network operators around the world are currently deploying 3rd generation (3G) networks, capable of delivering multimedia content. MPEG-4 Simple Profile has been selected by the Third generation Partnership Project¹ (3GPP) as the standard for wireless video in 3G mobile networks. The stability, interoperability and robustness offered by MPEG-4 makes it ideal for mobile networks.

The MPEG-4 Simple Profile was created with mobile visual services in mind amongst others. The Simple Profile at level 0 (Simple@L0) was specifically developed at the request of the 3GPP². The MPEG-4 Simple@L0 support rectangular natural video with frame rate of 15 fps in QCIF³ resolution and bit rates up to 64 Kbits/s.

Low bit rate multimedia transmission and exchange of MPEG-4 video content in a mobile network require mobile phones that are capable of handling highly compressed visual data. Consumption of low bit rate video requires mobile phones that are equipped with low-cost and low-power decoding capabilities.

¹ Third Generation Partnership Project: A group of telecommunication standard bodies to produce specification for the 3rd generation mobile system.

² Profiles and levels are describes in Section 2.4.1.

³ 144x176 pixels for luminance and 72x88pixels for chrominance.

This report presents a simulation model of a MPEG-4 Simple@L0 compliant video decoder in hardware for real-time decoding in mobile phones. The decoder has been modeled using behavioral modeling techniques.

Behavioral synthesis introduces a higher level of design abstraction and is defined as the transformation of a behavioral description to a Register Transfer Level (RTL) description. Synopsys tools were used for behavioral synthesis. These tools enable synthesis of designs, implemented at the behavioral level, and generate synthesizable code at the RTL level. The synthesis tools schedule operations into clock cycles and perform hardware allocation, based on user constraints. This allows designers to control design parameters such as latency, throughput, resource allocation and I/O activity without significant changes to the source code.

SystemC is a C++ class library and enables design and simulation of hardware descriptions in a software environment. SystemC allows designers to design and verify at all levels of abstractions in a common language. A manual conversion from an abstract architecture model in C++ to a detailed HDL RTL model can thus be avoided. The Synopsys Behavioral Compiler is a high-level synthesis tool capable of generating RTL code from a behavioral description. These tools show great promise in reducing time from concept to implementation by reducing design iterations and simulation times.

Hardware design in SystemC using behavioral design methodology offers several advantages. A software implementation of the algorithm, written in C++, can be gradually converted to a SystemC hardware model within the software environment. The model can, at any time, be verified for correctness in the software environment which often reduces simulation times. The higher level of abstraction introduced also promotes reuse.

Whether SystemC will be adopted as the future hardware modeling language remains to be seen. But currently SystemC offers an effective verification platform for high-level functional models because it introduces timing in a software environment. The success of behavioral synthesis will depend on the ability of synthesis tools to generate efficient design from behavioral descriptions.

An open source project known as XviD served as the basis for this project. The XviD CODEC is implemented in C making SystemC an obvious candidate as the implementation language. SystemC is a C++ class library which means that common C++ programming structures and data types, which can be simulated in the software environment, are not necessarily synthesizable. A major challenge of this project was to eliminate these structures from the XviD CODEC and rewrite the code to synthesizable alternatives. A lot of effort was put into identifying parts of the decoding process that can be processed concurrently, to make sure that these start processing as soon as input data is present. Additionally all data storage elements where carefully analyzed to reduce memory usage. This project uses version 0.91 of the XviD CODEC, which was the latest release at the start of this project. All future references to XviD in this report refer specifically to this version.

A behavioral model of a MPEG-4 Simple@L0 compliant video decoder, modeled in SystemC, and capable of decoding a MPEG-4 bitstream in real-time has been developed.

The model has been tested and verified for correctness at the behavioral level and only a few modules have been synthesized.

Chapter 2 presents the basics of MPEG-4 video. Chapter 3 describes those parts of the MPEG-4 video decoding process relevant to this project including some of the new tools introduced by the MPEG-4 standard. Chapter 4 includes a full description of the implemented model. Chapter 5 describes the testing methods used and some test results. A conclusion and ideas on future work of the model are included in Chapter 6.

A short description of the history and objectives of the MPEG-4 standard and the XviD project is included in Appendix A. The encoder settings used to generate the bitstream

used for testing is included in Appendix B. The data elements in the VOP header and the MB header are listed in Appendix C. The BMP file format is described in Appendix D. The AVI file format is briefly discussed in Appendix E. A list of all source files and other files and a description of how to compile and test the model can be found in Appendix F.

2 Basics of MPEG-4 Digital Video

This Chapter introduces some of the basics of digital video. Section 2.1 describes the YUV and RGB colorspace. Section 2.2 introduces the terminology used in the MPEG-4 standard. Section 2.3 briefly describes the MPEG-4 encoding process including texture coding and motion estimation. The new MPEG-4 visual tools used to code natural video are described in Section 2.4.

2.1 Colorspace

A colorspace defines how to represent color images. The YUV colorspace uses three components to describe an image. These are the Y component, called luminance, which describes a black and white image, and two other components U and V, which are the chrominance components and describe how to remove blue and red from the Y signal to create a color image. The term YCbCr¹ is often used when the three components are represented digitally and restricted to 8 bit integers.

The human eye perceives changes in brightness better than changes in color and therefore the color components are often sampled with less resolution than the luminance component. The format known as 4:2:0 uses half sampling of the chrominance components in the both the horizontal and the vertical direction. This is shown in Figure 2-1.

¹ The equivalent analogue term is YPbPr.

x o x	× ×	× × ×	× ×	× × ×	× ×	× o ×	× ×
x o x	××	× o ×	××	× o ×	××	× o ×	× ×
X O X	××	X O X	× ×	X O X	× ×	X O X	× ×

X Represent luminance samples

O Represent chrominance samples

Figure 2-1 The position of luminance and chrominance samples in 4:2:0 format.

(Source: Reference [1])

RGB colorspace uses three components the colors red green and blue. This colorspace is often used for displaying images on monitors. A conversion is often needed, when digital video has to be displayed on a monitor. The conversion formulas are shown below.

$$b = 1.164 (y-16) + 2.018 (u-128)$$

$$g = 1.164 (y-16) - 0.813 (v-128) - 0.391 (u-128)$$

$$r = 1.164 (y-16) + 1.596 (v-128)$$

Readers are referred to [4] for more information on colorspaces, formats and conversion formulas.

2.2 MPEG-4 Terminology

A frame is defined as three rectangular matrices of 8-bit integers; a luminance matrix (Y) and two chrominance matrices (Cb and Cr). A Video Object Plane (VOP) is obtained by

decoding a coded VOP. A coded VOP is derived from either a progressive or interlaced frame.

There are four types of VOPs and each type is coded differently. The Simple Profile supports Intra-coded (I) VOPs and the Predictive-coded (P) VOPs¹. I-VOPs are intra coded independently from other VOPs. P-VOPs are coded using a previously coded VOP called a reference VOP. The Simple Profile does not support B-VOPs so the decoding order is equivalent to the displaying order. A set of successive VOPs can be clustered in order to form a group of VOPs (GOV). This is useful for random access and resynchronization.

A macroblock (MB) is a 16x16 section of the luminance component and the spatially corresponding chrominance components. A MB consists of four 8x8 luminance blocks and two 8x8 chrominance blocks. These are ordered as illustrated in Figure 2-2.



Figure 2-2 4:2:0 Macroblock structure

The term block is used for an 8x8 matrix section of pixel values or DCT coefficients taken from one component. A MB consists of 6 blocks. The upper left DCT coefficient of each block is called the DC coefficient and the rest of the DCT coefficients in the block are called AC coefficients.

The MPEG-4 Simple Profile supports five types of MBs:

- 1. INTER (Used in P-VOP)
- 2. INTER+Q (Used in P-VOP)

¹ The other two types are Bidirectional (B) VOPs and Sprite (S) VOPs.

- 3. INTER4V (Used in P-VOP)
- 4. INTRA (Used in I-VOP and P-VOP)
- 5. INTRA+Q (Used in I-VOP and P-VOP)

INTER and INTER+Q macroblocks are inter-coded and only one motion vector is coded for the entire macroblock. INTER4V allows up to four motion vectors to be coded. INTRA and INTRA+Q are intra-coded macroblocks. INTER+Q and INTRA+Q macroblocks enable modification of the quantization stepsize. It is worth noting that a P-VOP may contain intra–coded macroblocks.

Progressive and interlaced VOPs are organized differently into MBs. For the case of a progressive VOP, the organization of luminance lines into MBs is called frame organization. For the case of an interlaced VOP, the organization of luminance lines into MBs can be either frame organization or field organization. Frame organization is illustrated in Figure 2-3 and field organization is illustrated in Figure 2-4.

Figure 2-3 Luminance macroblock structure in frame DCT coding

Source: Reference [1]



Figure 2-4 Luminance macroblock structure in field DCT coding

Source: Reference [1]

2.3 Basic MPEG-4 Encoding

This section gives a brief description of the MPEG-4 Simple Profile encoding process for natural video. Two types of VOPs are used. I-VOPs and P-VOPs. I-VOPs are coded completely independent from other VOPs, where as P-VOPs are coded using a previous VOP as reference. For the Simple Profile the reference VOP is always the last coded VOP. Because P-VOPs are coded using information already send they can often be much more efficiently coded, however using too many P-VOPs makes editing and access more difficult and the bitstream more susceptible to error.

P-VOPs are coded using motion estimation followed by texture coding of the residue. I-VOPs are coded using texture coding only.

Encoder settings can be found in Appendix B.

2.3.1 Motion estimation

Motion estimation is block based and involves searching in the reference VOP for the closest match to the current block to be coded. Once the block is found, a motion vector is used to describe its location. This is illustrated in Figure 2-5.



Figure 2-5 Motion Estimation

Motion estimation is performed on each of the four luminance blocks in a MB. Depending on the MB type up to four motion vectors are coded. Motion vectors for the two chrominance blocks are not coded but derived from the luminance motion vectors.

Half-pel motion vectors require interpolation to calculate half-pel values. Half-pel interpolation is described in Section 3.2.2.

2.3.2 Texture coding

For I-VOPs the actual coefficients are coded in the bitstream. For P-VOPs only the residue is coded. In both cases the coefficients are transformed from the spatial domain to the frequency domain using the Discrete Cosine Transform (DCT). The coefficients are then quantized to enable further compression. Quantization is the only part of the coding process that introduces deliberate quality loss. DCT concentrates the information in fewer coefficients and quantization introduces loss of quality in a way less noticeable to the

human eye. A full analysis of the theory behind this method is beyond the scope of this project.

Finally the quantized DCT coefficients are coded using Variable Length Codes (VLC). The data no longer have externally identifiable boundaries. The idea behind VLC is to assign shorter codes for the most common coefficients. VLC is described in Section 3.1.

2.4 MPEG-4

This section includes a description of the MPEG-4 tool set used for natural video. The history and objectives of the MPEG-4 standard and the XviD project can be found in Appendix A.

2.4.1 Profile and Levels

To ensure interoperability between MPEG-4 implementations, profiles and levels have been standardized. Different profiles have been created for different application areas to allow users to implement only a subset of the tools available in the MPEG-4 standard and still be compliant. Levels define the bounds of complexity for a particular profile, e.g. maximum bit-rate or spatial resolution. The most popular visual profiles are the Simple and the Advanced Simple Profile.

The Advanced Simple Profile is a superset of the Simple Profile and has added tools that enhance compression efficiency. These tools include quarter-pel motion-estimation, GMC (global motion estimation) and B-frames.

XviD v.0.91, which is the version of the XviD CODEC used in this project, supports the Simple Profile (current version supports the Advanced Simple Profile).

The Simple Profile was created for low-complexity applications. Application areas include mobile multimedia services, low bit rate video on the internet or recording of video on memory chips.

The Simple@L0 was defined specifically at the request of the 3GPP. Simple@L0 supports a maximum bit rate of 64Kbits/s, a maximum image resolution of QCIF and a maximum frame rate of 15 fps.

For more information on profiles and levels readers are referred to [1] and [3].

2.4.2 Natural Video

Both the Simple Profile and the Advanced Simple Profile define a tool set very similar to those used in earlier MPEG standards. Coding of rectangular natural video still uses the conventional block-based hybrid coding scheme, but with new and improved tools.

New motion compensation tools

- 1. **Quarter-pel motion-compensation**: Previous standards such as H.261, H.263, MPEG-1 video and MPEG-2 video used only half-pel resolution motion vectors.
- 2. Global Motion Vector (GMC): A single motion vector is coded for the entire frame. Useful for coding sequences with large global motion, e.g a moving camera.
- 3. **Direct mode in bidirectional prediction**: Bidirectional motion-compensated prediction using motion vectors of neighboring P-frames.

New texture coding tools

- 1. **Quantization of DCT transform Coefficients**: Two quantization procedures can be applied. The first is derived from the MPEG-2 video standard and an additional procedure used in H.263.
- 2. AC/DC Prediction for Intra Macroblocks: Statistical dependencies between neighboring macroblocks are exploited to predict a coefficient of one block from a coefficient in a neighboring block.

3. Alternative Scan Modes: Previous standards used the conventional zig-zag scan mode. MPEG-4 video introduces two additional scan modes. The Alternate vertical scan mode and the alternate horizontal scan mode.

The new tools are all included in the Advanced Simple Profile.

Simple@L0 supports AC/DC prediction, alternative scan modes, 4-motion vectors and unrestricted motion vectors. B-VOPS, MPEG quantization, quarter-pel motion compensation and GMC are not supported by Simple@L0.

A full description of all the tools supported by Simple@L0 will be given in Chapter 3.

2.5 Summary

The YUV colorspace uses a luminance and two chrominance components to represent color images. The 4:2:0 format uses sub sampling of the chrominance components. Often images needs to be converted to the RGB colorspace to be displayed on a monitor.

MPEG-4 terminology includes VOP, I-VOP, P-VOP, GOV, macroblock, block, progressive, interlaced, frame and field organization.

Encoding involves motion estimation and texture coding.

MPEG-4 standard includes new tools for coding rectangular natural video which improves compression ratio.

3 MPEG-4 Video Decoding Process

This Chapter describes the decoding process. Only the parts relevant to this project, i.e. those included in the Simple@L0 are described.

The description of the decoding process and the bit stream format has been simplified and readers are referred to [1] for a more detailed description. Except for the IDCT the decoding process is defined such that all decoders shall produce numerically identical results. Figure 3-1 shows a simplified diagram of the decoding process.



Figure 3-1Simplified Video Decoding Process

Source: Reference [1]

Decoding of natural video is composed of a texture decoding part and a motion decoding part. In case of intra-coded MBs no motion information is coded, and the MB is reconstructed from the decoded texture pixel values. In case of inter-coded MBs the decoded motion and texture pixel values are added to form the MB. In both cases pixel values are saturated to the interval [0; 255]. Each coded VOP includes a VOP header followed by the coded MBs. Each coded MB contains a MB header and depending on its type, motion data and texture data.

The VOP header and MB header data elements are listed in Appendix C.

3.1 Texture Decoding

Each block of DCT coefficients are coded as a sequence of EVENTS. Each EVENT represents a non-zero coefficient in the block. An EVENT is a combination of (LAST, RUN, LEVEL), where LAST is a 1 bit value indicating, whether there are more non-zero coefficients in this block. RUN indicates the number of consecutive zeros preceding the DCT coefficient. LEVEL is the magnitude of the DCT coefficient.

Three different scan patterns are used to convert the sequence of decoded coefficients into a two-dimensional block. The scan-patterns are shown in Figure 3-2.

0	1	2	3	10	11	12	13
4	5	8	9	17	16	15	14
6	7	19	18	26	27	28	29
20	21	24	25	30	31	32	33
22	23	34	35	42	43	44	45
36	37	40	41	46	47	48	49
38	39	50	51	56	57	58	59
52	53	54	55	60	61	62	63

0	4	6	20	22	36	38	52
1	5	7	21	23	37	39	53
2	8	19	24	34	40	50	54
3	9	18	25	35	41	51	55
10	17	26	30	42	46	56	60
11	16	27	31	43	47	57	61
12	15	28	32	44	48	58	62
13	14	29	33	45	49	59	63

0	1	5	6	14	15	27	28
2	4	7	13	16	26	29	42
3	8	12	17	25	30	41	43
9	11	18	24	31	40	44	53
10	19	23	32	39	45	52	54
20	22	33	38	46	51	55	60
21	34	37	47	50	56	59	61
35	36	48	49	57	58	62	63

Figure 3-2 (a) Alternate horizontal scan (b) Alternate vertical scan (c) ZigZag scan

The sequence of EVENTs is coded in the bitstream using VLCs. The statistically most common EVENTs are assigned a predefined VLC from a VLC code table. Different code tables are used for inter and intra blocks. Figure 3-3 shows part of the VLC code table for intra blocks.

VLC CODE	LAST	RUN	LEVEL
10s	0	0	1
1111 s	0	0	3
0101 01s	0	0	6
0010 111s	0	0	9
0001 1111 s	0	0	10
0001 0010 1s	0	0	13

Figure 3-3 Part of the VLC Table for Intra Blocks

Source: Reference [1]

S is the sign bit, i.e. the next bit in the bitstream. VLCs are used to code the sequence of EVENTS more efficiently. The idea is to use shorter VLCs for the most common EVENTS. Many EVENTs are not assigned a VLC. An escape code method is used to encode these statistically rare EVENTs. Escape codes are described in [1].

The DC coefficient of a block in an intra-coded MB is encoded differentially using VLC. This differential DC value is added to a prediction value to get the final DC value. The DC prediction value is obtained from a neighboring block and is determined by the DC prediction direction. This process will be described in Section 3.1.1.

 $DC = DC_{predicted} + DC_{differential} / dc _ scaler$

The *dc_scaler* is obtained from the MB header. All other coefficients are obtained by decoding the VLCs from the bitstream.

For intra blocks if $acpred_flag^1 = 0$ the zigzag scan is used. Otherwise DC prediction direction is used to select a scan. All other blocks use the zigzag scan.

It is worth noting that a sequence of VLCs have no externally identifiable boundaries.

¹ acpred_flag is a 1-bit value coded in the VOP header.

3.1.1 DC and AC Prediction Direction

DC prediction is only carried out for intra coded macroblocks. The DC prediction direction is based on a comparison of the horizontal and vertical DC gradients around the block to be coded. Figure 3-4 illustrates the method.



Figure 3-4 DC Prediction Direction



The prediction direction of block X is determined as follows.

 $if \left(\left| DC(A) - DC(B) \right| < \left| DC(B) - DC(C) \right| \right)$ $predict \ from \ block \ C$ else $predict \ from \ block \ A$

The prediction block found during DC prediction is also used for AC prediction, when $acpred_flag = 1$. AC prediction involves using the coefficients from the first row or the first column of the prediction block to predict the co-sited AC coefficients in the current block. The predicted coefficients are added to the coefficients coded from the bitstream to get the final AC coefficients. This is illustrated in Figure 3-5.



Figure 3-5 AC Prediction
Source: Reference [1]

X is the current block. If block A was selected as the predictor the first column of block A holds the AC predictors. If block C was selected as the predictor the first row of block C holds the AC predictors. The predictors are scaled by the ratio of the quantization stepsize for the current and the predicted block before being added to their collocated coefficients.

$$AC_{i,j(current)} = AC_{i,j(differential)} + AC_{i,j(predicted)} * quant_stepsize_{predicted} / quant_stepsize_{current}$$

 $quant_stepsize_{predicted}$ is the quantization stepsize of the prediction block. $quant_stepsize_{current}$ is the quantization stepsize of the current block. AC_{i,j(predicted)} is the predicted coefficient, $AC_{i,j(differential)}$ is the coded coefficient and AC_{i,j(current)} is the final coefficient.

A quantization stepsize is defined for every MB and is derived from the VOP header and the MB header. The quantization stepsize is equivalent to *vop_quant¹*. *vop_quant* may

¹ *vop_quant* is a 5-bit unsigned integer coded in the VOP header

sub sequentially be changed by *dquant¹*. Changes made to *vop_quant* are permanent and valid for all subsequent MBs.

DC and AC prediction allow the first row or column of each block to be coded differentially and enables these coefficients to be coded more efficiently as numerically smaller coefficients are assigned shorter VLCs.

The actual prediction process includes boundary checking, but the basic principles of DC and AC prediction have been illustrated.

When DC and AC prediction is used the quantization stepsize and the predictors must be stored for the two most recently coded rows of macroblocks, before inverse quantization is performed.

The final step before inverse quantization is to saturate the coefficients to lie in the interval range [-2048; 2047].

3.1.2 Inverse Quantization

Inverse quantization reconstructs the original DCT coefficients and involves multiplication with the MB quantization stepsize. If $quant_type^2 = 1$ the first method is used else the second method is used.

3.1.2.1 First Inverse Quantization Method

Two weighting matrices are used in this process. A weighting matrix W is an 8x8 block of 8-bit unsigned integers. One is used for intra-coded MBs, and the other for inter-coded MBs. Each matrix has a set of default values that may be overwritten by user defined values. The default weighting matrices are shown in Figure 3-6.

¹ dquant is 2-bit code in the MB header and specifies changes to vop_quant

² quant_type is a 1 bit value coded in the VOP header.
Figure 3-6 a) Default matrix for intra blocks b) Default matrix for inter blocks

The DC coefficient of intra coded blocks is inverse quantized by multiplying with the dc_scaler .

$$F_{0,0} = Q_{0,0} \times dc _ scaler$$

All other coefficients are inverse quantized using the below equation.

$$F_{i,j} = \begin{cases} 0, & \text{if } Q_{i,j} = 0\\ ((2 \times Q_{i,j} + k) \times W_{i,j} \times quant_stepsize)/16, & \text{if } Q_{i,j} \neq 0\\ & \text{where} \end{cases}$$
$$k = \begin{cases} 0 & \text{intra blocks}\\ Sign(Q_{i,j}) & \text{inter blocks} \end{cases}$$

 $F_{i,j}$ is the inverse quantized DCT coefficients. $Q_{i,j}$ is the original coefficients and $W_{i,j}$ is the weighting matrix.

3.1.2.2 Second Inverse Quantization Method

DC coefficients of intra coded blocks are quantized using the same method as described in Section 3.1.2.1. All other coefficients are inverse quantized using the below equation.

$$|F_{i,j}| = \begin{cases} 0, & \text{if } Q_{i,j} = 0\\ (2 \times |Q_{i,j}| + 1) \times quant_stepsize, & \text{if } Q_{i,j} \neq 0, \text{ quant_stepsize is odd}\\ (2 \times |Q_{i,j}| + 1) \times quant_stepsize - 1, & \text{if } Q_{i,j} \neq 0, \text{ quant_stepsize is even} \end{cases}$$

 $F_{i,j}$ is the inverse quantized DCT coefficients. $Q_{i,j}$ is the original coefficients and *quant_stepsize* is the quantization stepsize of the current MB. The sign of $F_{i,j}$ is equivalent to the sign of $Q_{i,j}$.

3.1.2.3 Saturation

The coefficients resulting from the inverse quantization process are saturated to lie in the interval [-2048; 2047].

3.1.3 Inverse DCT

The inverse DCT converts the inverse quantized coefficients from the frequency domain to the spatial domain.

The 8x8 2D DCT is defined as

$$F(u,v) = \frac{1}{4}C(u)C(v)\sum_{x=0}^{7}\sum_{y=0}^{7}f(x,y)\cos\frac{(2x+1)u\pi}{16}\cos\frac{(2y+1)v\pi}{16}$$

where
$$C(u), C(v) = \begin{cases} \frac{1}{\sqrt{2}} & \text{for } u, v = 0 \end{cases}$$

f(x,y) are the original 8 bit pixel values and F(u,v) are the DCT coefficients.

The 8x8 2D inverse DCT (IDCT) is defined as

$$f(x, y) = \frac{1}{4} \sum_{u=0}^{7} \sum_{v=0}^{7} C(u)C(v)F(u, v)\cos\frac{(2x+1)u\pi}{16}\cos\frac{(2y+1)v\pi}{16}$$

It is worth noting that implementing the 8x8 IDCT as defined above requires 64x64=4096 iterations.

3.2 Motion Decoding

Motion decoding is performed on all inter-coded MBs.

3.2.1 Padding

In order to perform motion compensation the region outside of the reference VOP must be padded using a special padding process. The padding process is a two step process. Horizontal repetitive padding is performed by replicating the edge samples to the left and right direction in order to fill the regions. Similarly a vertical repetitive padding is performed to fill the top and bottom regions. Figure 3-7 illustrates the padding process of an 8x8 frame with a padded region size of three.



Figure 3-7 Padding

Padding is performed to support unrestricted motion compensation as described in Section 3.2.3.

3.2.2 Half sample interpolation

Half sample interpolation is used when motion vectors in one or both directions are halfpel resolution. The interpolation process samples values in between the actual pixels.



Figure 3-8 half-pel Interpolation



The half-pel samples shown in Figure 3-8 are calculated using the following equations

a = A $b = (A + B + 1 - rounding _ type)/2$ $c = (A + C + 1 - rounding _ type)/2$ $d = (A + B + C + D + 2 - rounding _ type)/4$

rounding_type is a 1-bit unsigned integer coded in the VOP header. Only sample values inside the padded reference VOP may be used for interpolation.

3.2.3 Unrestricted motion compensation

When unrestricted motion compensation is supported, a sample referenced by a motion vector, may lie outside the reference VOP. In this case an edge sample is returned instead. This process is done on a sample basis by limiting the horizontal and the vertical component of the motion vector. This is illustrated in Figure 3-9.



Figure 3-9 Unrestricted Motion Compensation

Source: Reference [1]

The coordinates of the reference sample are calculated using the following equations

$$xref = MIN(MAX(Xcur + dx, 0), width - 1)$$

$$yref = MIN(MAX(Ycur + dy, 0), height - 1)$$

(*Xref, Yref*) is the coordinate of the reference sample, (*Xcur, Ycur*) is the coordinate of the current sample, and (dx, dy) is the motion vector. *width* and *height* are the VOP dimensions.

Basically (Xref, Yref) is limited to only reference samples inside the reference VOP. The padding method is used to implement unrestricted motion compensation. Padding extends the region outside the reference VOP with edge samples.

The Simple@L0 limits the components of the motion vector to the range [-32, 31] and unrestricted motion compensation can be implemented by extending the reference VOP with a padding region which is 32 sample wide for luminance and a 16 sample wide for chrominance.

3.2.4 Vector decoding

The differential motion vectors are extracted from the bitstream and added to a prediction to form the final motion vector. Vector prediction is describes in Section 3.2.5. A differential motion vector (MVDx, MVDy) is variable length coded. This is shown next for the horizontal component.

$$r_size = vop_fcode_forward -1;$$

$$f = 1 << r_size;$$

$$if((f == 1) \parallel (horizontal_mv_data == 0))$$

$$MVDx = horizontal_mv_data;$$

$$else\{$$

$$MVDx = ((ABS(horizontal_mv_data) - 1)*f) + horizontal_mv_residual + 1$$

$$if(horizontal_mv_data < 0)$$

$$MVDx = -MVDx;$$
}

 $vop_fcode_forward$ is a 3-bit unsigned integer coded in the VOP header and is used to decode motion vectors. *horizontal_mv_data* is a variable length coded integer in the range [-32; 31] and *horizontal_mv_residual* is an unsigned integer, which is coded using *r_size* bits. *horizontal_mv_residual* is only coded if $vop_fcode_forward > 1$.

vop_fcode_forward is also used to restrict the final motion vectors as shown in Table 3-1.

vop_fcode_forward	motion vector range			
	(half sample units)			
1	[-32; 31]			
2	[-64; 63]			
3	[-128; 127]			
4	[-256; 255]			
5	[-512; 511]			
6	[-1024; 1023]			
7	[-2048; 2047]			

Table 3-1 Range of Motion Vectors

Source: Reference [1]

For Simple@L0 *vop_fcode_forward* is always 1, which implies that the *mv_residual* fraction is not coded in the bitstream. This simplifies decoding of differential motion vectors to decoding of a VLC. Additionally motion vectors are restricted to the range [-32; 31] ensuring that the padding process described above implements unrestricted motion compensation.

3.2.5 Vector prediction

A prediction MV is formed using three vector candidate predictors MV1, MV2 and MV3 for each of the four luminance blocks. These candidate predictors are taken from blocks in the spatial neighborhood of the current block. This is illustrated in Figure 3-10.



Figure 3-10 Candidate Predictors MV1, MV2, MV3 for each of the luminance blocks

Source: Reference [1]

A candidate is valid if the corresponding block is within the boundary of the current video packet. Video packets are described in Section 3.3.

A single prediction vector is formed from the candidates using the following process.

- 1. If one and only one candidate is not valid it is set to zero.
- 2. If two and only two candidates are not valid, they are set to the third candidate.
- 3. If three candidates are not valid, they are set to zero.
- 4. The median value of the three candidates is computed as the predictor.

The predictors are added to the corresponding differentially coded vectors to form the final motion vectors for the luminance blocks. The motion vector for both chrominance blocks is derived by calculating the average of the luminance vectors.

The actual process involves excluding luminance blocks that are not coded from the calculations.

3.3 Error resilience

MPEG-4 has adopted error resilience tools that provide basic error robustness. The following tools are specified for the Simple@L0.

- Slice resynchronization
- Data partitioning
- Reversible VLC

These tools facilitate resynchronization, error localization, data recovery, and error concealment. However the MPEG-4 standard does not specify which actions the decoder should take, when an error is detected, although some strategies are suggested in Annex E of the MPEG-4 standard [1].

3.3.1 Slice resynchronization

Slice resynchronization attempts to reestablish synchronization between the decoder and the bitstream after an error has been detected. This is achieved by inserting resync markers in the bitstream. A resync marker is a unique code that cannot be emulated by the encoder.

When an error is detected the decoder finds the next resync marker and synchronization is established. MPEG-4 has adopted a packet-based resynchronization solution. Each video packet consists of an integer number of consecutive coded MBs preceded by a video packet header. The video packet header contains a resync marker followed by duplicated VOP header information. Video packets enable the decoder to localize errors to within a few MBs. In addition to inserting resync markers all dependencies between consecutive video packets are eliminated. This ensures that a corrupt video packet does not hinder the other video packets from being decoded. Prediction tools such as AC/DC prediction and motion vector prediction are limited by video packet boundaries. This limits error propagation.

3.3.2 Data partitioning

Data partitioning has been adopted to enable better resynchronization and error localization. Data partitioning separates the MB data into high-priority and low-priority components. For intra-coded MBs the DC coefficients are separated from the AC coefficients and a DC marker (DCM) is inserted between the two parts. For inter-coded MBs the motion data is separated from the texture data and a motion marker (MM) is inserted between the two parts.

3.3.3 Reversible VLC

Using reverse VLC (RVLC) to encode the DCT coefficients enables the decoder to better isolate errors. RVLCs are special VLCs that can be uniquely decoded in the forward and reverse direction.

When an error is detected the next resync marker is found, and from there the bitstream is decoded, in the backward direction, until a new error is detected. Based on the location of the two errors, the decoder can recover some of the data between the two synchronization points.

The MPEG-4 standard includes an efficient and robust RVLC table for encoding DCT coefficients. However these codes are only used in conjunction with data partitioning.

3.4 Simple@L0

The Simple@L0 includes the following features and tools.

- Decoding of progressive rectangular video in the 4:2:0 format.
- A maximum resolution of QCIF.
- A maximum bit rate of 64 Kbits/s.
- A maximum frame rate of 15 fps.
- I-VOPs and P-VOPs.
- Quantization method 2 (H.263).
- AC/DC prediction.
- Motion vector prediction.
- Alternative scan modes.
- 4-motion vectors.
- Unrestricted motion compensation.
- Short video header.
- Error resilience
 - Slice resynchronization
 - Data partitioning
 - o RVLC

3.5 Summary

This chapter presented a brief overview of the MPEG-4 Simple@L0 decoding process for rectangular natural video.

Texture decoding involves decoding a sequence of VLCs with no externally identifiable boundaries. Each VLC represents a non-zero coefficient and the texture blocks are reconstructed using a predefined code table and a pre-selected scan pattern. AC/DC predictors are added before the blocks are inverse quantized and IDCT.

Motion decoding involves using motion vectors to do half-pel interpolation. Motion vectors are obtained by decoding differential motion vectors from the bitstream and

adding predictors. The reference VOP is padded to support unrestricted motion compensation.

The MPEG-4 standard has adopted several error resilience tools. These include slice resynchronization, data partitioning, and RVLC.

Simple@L0 uses simple coding tools based on I-VOPs and P-VOPs to code rectangular natural video.

4 Implementation

This chapter describes the implementation of the MPEG-4 Simple@L0 compliant decoder. The implemented decoder supports all tools specified by the Simple@L0 except for data partitioning, RVLC, and short video header. Bit errors are handled by the slice resynchronization tool, which discards all macroblocks in the video packet containing the error.

4.1 Behavioral Modeling

This section describes the process of converting the XviD CODEC to a synthesizable hardware model in SystemC.

The first step was to separate the decoder and from the encoder. This included identifying which data structures and functions where used in the decoder and which where used in the encoder. Next the internal structure of the decoder was determined and specified as separate modules.

Then the internal communication between modules was specified. This involved defining communication ports for each module, and whether to use dedicated or shared communication resources, such as a bus, for inter-module communication. At this point the communication protocol was designed for all communication resources.

Then the internal structure of each module was specified. The necessary data structures and processes were declared. Processes within a module are concurrent and enable parallel behavior of hardware to be modeled. SystemC provides three types of processes, however only the clocked thread process SC_CTHREAD is supported for behavioral synthesis. A clocked thread process is only sensitive to one edge of the clock and models the behavior of sequential circuits with unregistered inputs and registered outputs. Consequently behavioral synthesis does not allow combinatorial logic to be modeled.

Internal signals were specified for inter-process communication. Internal signals are always registered. Inter-process communication using variables is not allowed.

Finally the behavior of the modules was modeled. Reset behavior were determined and implemented and followed by an implementation of processes.

4.2 Converting to Synthesizable Subset of SystemC

Many C and C++ language constructs and SystemC classes cannot be synthesized into hardware. This section describes how these constructs have been converted to a synthesizable subset.

The following list contain some of the non-synthesizable construct found in the XviD CODEC and describes how they were corrected.

- **Dynamic storage allocation**: malloc, free, new, delete, etc. were replaced with static memory allocation.
- **Recursive function call**: These were replaced with iterations.
- C++ built-in functions: The math library and similar built-in functions were implemented manually and I/O library functions were commented out.
- Dereference operator: * and & operators were replaced with direct access to variables and arrays.
- Sizeof operator: Size was determined statically.
- **Pointer**: Pointers were replaced with direct access to variables and arrays.
- **Reference**: & were replaced with direct access.
- **Type casting at run-time**: Data types were used that would require no type conversions.

For a full list of all non-synthesizable C/C++ constructs and SystemC classes, readers are referred to [2].

The following list contains SystemC and C++ data types that are not synthesizable.

- Floating-point types such as *float* and *double*.
- Fixed-point types: *sc_fixed*, *sc_ufixed*, *sc_fix*, *sc_ufix*.
- Access types: pointers.
- File types: *FILE*.
- I/O streams: *stdout* and *cout* are ignored by the synthesis tools.
- *Sc_logic* and *sc_lv*: Used only for RTL synthesis.

Behavioral synthesis does not support four-value logic signals, which means that tri-state signals cannot be modeled.

It is important to use appropriate bit-widths so that the synthesis tools do not build unnecessary hardware. C/C++ models typically uses native C/C++ types such as *int*, *char*, *bool* or *long*. These types have platform-dependent widths, which are often not optimal for hardware.

All native data types were evaluated and converted to data types of appropriate widths.

The *wait* statement suspends process execution for one clock cycle. The following list contains the six general coding rules for behavioral synthesis.

- 1. Place at least one *wait* statement in every loop.
- 2. Place at least one *wait* statement between successive writes to the same signal.
- 3. Place at least one *wait* statement before the main loop.
- 4. If one branch of a conditional has at least one *wait* statement, place at least one *wait* statement in each of the other branches (balancing).
- 5. Place at least one *wait* statement after the last write inside a loop and before a loop continue.
- 6. Place at least one *wait* statement after the last write before a loop.

4.2.1 Pipelining Loops

By default loops are not pipelined. Pipelining loops can significantly increase throughput and overall runtime latency. Basically a pipelined loop is synthesized so that loop iterations overlap. Initiation interval and loop latency must be specified to pipeline a loop. The initiation interval is the number of clock cycles until the start of the next loop iteration, and loop latency is the number of clock cycles required to complete one iteration. The loop latency must be an integer multiple of the initiation interval. Throughput is a function of the initiation interval specified. The initiation interval is determined by careful analysis of

- Loop carry dependencies: Data produced in one iteration of a loop and consumed in a subsequent iteration.
- Memory and I/O accesses: Reading and writing to the same memory, signal or port are not possible from different loop iterations.
- Handshake signals: The order in which handshake signal are raised and lowered cannot be changed.
- Exit from loop: A loop exit must occur within the initiation interval to prevent inadvertent launching of future iterations.

A pipelined loop cannot contain a structure that implies waiting for some external event to occur (such as waiting for an input signal to be raised) as this would imply a runtime determined loop latency and require stalling the pipeline. Consequently two-way handshake communication is not allowed inside a pipelined loop.

Modules that are critical to overall latency have been implemented with pipelining in mind.

4.2.2 Signals

Signals are means by which processes and modules can communicate. Inter-process communication through shared variables is not synthesizable because it can lead to

indeterminism. All signals are registered and written on the rising edge of the clock. All boolean signals are asserted high unless otherwise specified.

4.3 Structure

A diagram illustrating the internal structure of the decoder is shown in Figure 4-1.



Figure 4-1 Internal structure

The module *getBits* extract an arbitrary (\leq 32) number of bits from the bitstream. *Demux* outputs the control signals from the decoded VOP header and separates the texture and motion data. Texture decoding and motion decoding are described in Section 4.3.1 and Section 4.3.2. The *Reconstruct* module collects texture and motion MBs and stores the reconstructed MBs in the *Frame buffer*. The *Memory Controller* controls access to the *Frame buffer* and the *Reference VOP*. The *Display* module converts the reconstructed VOPs in the *Frame buffer* to RGB values. The *Output Generator* sends the RGB values to a monitor line by line.

Texture and motion decoding is done on a MB basis. Intra-coded MBs involves only texture decoding. Not-coded MBs involves only motion decoding, i.e. the MB is copied

directly from the reference VOP. Inter-coded MBs involves both texture and motion decoding.

A long sequence of consecutive intra or not-coded MBs may cause uneven workload of texture and motion decoding parts. Therefore first in first out (FIFO) buffers has been inserted to enable the *Demux* module to continue reading from the bitstream, while previous MBs are being processed.

The decoder is said to be in an idle state when it is not processing any MBs.

4.3.1 Texture decoding

The internal structure of texture decoding is shown in Figure 4-2.



Figure 4-2 Internal structure of Texture decoding

The *Scan Prediction* module performs VLC decoding and AC/DC prediction, using the proper scan mode, to build MBs. *Tbuffer* is a FIFO buffer and stores up to four MBs. *Quant* performs inverse quantization on a MB and *Idct* performs the Inverse Discrete Cosine Transform function on a MB. The structure in Figure 4-2 will also be referred to as the texture pipeline.

4.3.2 Motion decoding

The internal structure of motion decoding is shown in Figure 4-3.



Figure 4-3 Internal structure of Motion decoding

MV Prediction decodes the motion vectors and calculates the prediction candidates. The decoded motion vectors are stored in *Mbuffer*, which is a FIFO buffer that stores up to four elements of four motion vectors. The *Interpolate* module uses the reference VOP to perform interpolation. The structure in Figure 4-3 will also be referred to as the motion pipeline.

4.4 Communication

In this section the communication protocol used for inter module communication is described. Behavioral synthesis can insert additional clock cycles in order to schedule a design. Therefore, a design verified at the behavioral level may no longer work once scheduled to RTL level. Handshaking ensures proper behavior of the communication between modules after scheduling. Two-way handshaking ensures correct data transfer, when modules do not have a fixed response time. The implemented decoder uses two-way handshake protocols for all inter module communication. Figure 4-4 shows the handshake signals used for the two-way handshake protocol.



Figure 4-4 Receiver-initiated two-way handshake signals

The timing diagram for the two-way handshake protocol is shown in Figure 4-5.



Figure 4-5 Receiver-initiated two-way handshake protocol

The two-way handshake protocol has a minimum delay of one clock cycle.

In cases where multiple data has to be transferred the two-way handshake protocol is used to initiate the transfer. The timing diagram for the two-way handshake transfer of multiple data is shown in Figure 4-6.



Figure 4-6 Two-way handshake transfer of multiple data

In this case both participating modules must complete the transfer once it has been initiated. This protocol has been used for transferring MBs between modules. A MB uses 6*64=384 clock cycles to complete a transfer.

The two types of two-way handshake protocols have been used to initiate all data transfers between modules. All future references to handshake protocols refer to the two-way handshake protocol, unless otherwise specified.

4.5 Input/Output FIFO buffers

All image processing modules are equipped with input and output buffers. This is shown in Figure 4-7.



Figure 4-7 Input/Output buffers

Each buffer can store data from two complete MBs. Buffers are inserted so that the processing units do not spend time clocking in and out data. As long as there are unprocessed MBs in the input buffer and available space in the output buffer, processing of the next MB can continue with no additional latency. Filling and emptying the buffers are handled by two separate processes that control the interface to external modules.

4.6 Modules

The following section describes each of the implemented modules. The interface and processes of each module is shown in diagrams. All modules have a *clk* and a global *reset* input port, but these are not shown in the diagrams. The format of the diagrams is shown in Figure 4-8.



Figure 4-8 Format of Diagrams

The grey box indicates the module boundary. Signals between processes are internal signals used for inter-process communication. Input and output ports are used for communication with other modules.

All source files are listed in Appendix F with a reference to the corresponding module.

4.6.1 getBits

getBits enables a specific number (\leq 32) of bits to be read from the bitstream. The module reads the bitstream from the memory and sends the requested amount of bits to its output. The module is shown in Figure 4-9.



Figure 4-9 getBits

An internal buffer of eight bytes is continuously filled with bits from the bitstream. The bitstream is read from the buffer using the handshake signals *req_bits* and *ready_bits*. *num_bits* determines the number of bits to read and *bits* is the 32-bit output. In case *show_bits* is asserted the bitstream position is not updated after the current read. In case *from_byte_align* is asserted the current read is byte aligned. In case *num_to_byte_aligned* is asserted the number of bits to the next byte aligned position is returned. All signals are asserted high.

The module enables a specific number of bits from the bitstream to be read. This can be done without updating the bitstream position and byte aligned. Finally the offset to the next byte aligned position can be read.

4.6.2 Demux

Demux separates the VOP header, the texture data and the motion data in the bitstream in preparation for texture and motion decoding. The module is shown in Figure 4-10.



Figure 4-10 Demux

4.6.2.1 Header

The VOP header is read from the bitstream and written to the control bus as soon as the previous VOP has been decoded. All control signals are driven by the *Demux* module. The control signals are listed in the Table 4-1.

Name	Width	Comments			
short_video_header	1	Always 0 (short_video_header not supported)			
frame_type	3	Either 0 (I-VOP) or 1 (P-VOP)			
quant_type	1	Determines which inverse quantization			
		method to be used			
interlacing	1	Always 0 (Interlacing not supported)			
edged_width*	14	VOP width including padding			
edged_height*	14	VOP height including padding			
mb_width*	10	VOP width in MB units			

mb_height*	10	VOP height in MB units
width*	13	VOP width
height*	13	VOP height
rounding	1	Used for rounding in interpolation
intra_dc_threshold	6	Used to decode DC coefficient
quant	10	VOP quantization stepsize
fcode_forward	3	Used to decode motion vectors

Table 4-1 VOP header

*These values refer to luminance

Table 4-2 shows where the control signals are used.

Modules Control Signals	Scan Prediction	Quant	Idct	MV Prediction	Interpolation	Reconstruct	Memory Controller	Display	Output Generator
short_video_header	•								
frame_type						•			
quant_type		•							
interlacing				•		•			
edged_width					•	•	•		
edged_height					•	•	•		
mb_width	•			•	•	•	•		
mb_height					•	•	•		
width						•	•	●	•
height						•	•	٠	•
rounding					•				
intra_dc_threshold	•								
quant	•								
fcode_forward				•					

Table 4-2 Control signals versus modules

short_video_header and *interlacing* are assumed to be zero but written to the control bus to ease future extensions to the supported toolset.

The *matrix* group of signals is used to update the weighting matrices, when inverse quantization method one is used. The matrices are coded in the bitstream as part of the VOP header. Handshaking is used to communicate with the *Quant* module.

The *update_ref* group of signals is used to initiate an update of the reference VOP. The signals use handshaking to communicate with the *Memory Controller* to ensure that the decoder is idle, i.e. the previous VOP has been decoded, before the reference VOP is updated. The control bus and the matrices are only updated, when the decoder is idle. The *Memory Controller* completes the handshake communication with the *Demux* module immediately after the last MB has been written to the frame buffer. This ensures that the update of the reference VOP is initiated as soon as possible. Thus avoiding unnecessary delay for the interpolate module.

The update of the reference VOP is controlled by the *Memory Controller* enabling the *Demux* module to continue decoding the bitstream. The decoder can commence decoding the next VOP as soon as the control signals and the matrices are updated, keeping the decoder in the idle state as short as possible. The motion pipeline is stalled, while the reference VOP is being updated, however the use of internal buffers will keep the texture pipeline busy. The texture pipeline, which includes the *Idct* module, will under normal circumstances have a greater latency than the motion pipeline.

If the next VOP is intra-coded the *update_ref* group of signals is used only to detect the idle state. The reference VOP is in this case not updated.

4.6.2.2 Motion

The differential motion vectors are decoded from the bitstream and send to the *MV Prediction* module using the *motion_vector0-3* group of signals. Depending on the coded mode of the MB up to four motion vectors may be send. The *motion_header* group of signals is decoded from the MB header and is listed in Table 4-3.

Name	Width	Comments
Х	10	Horizontal position of MB in VOP
у	10	Vertical position of MB in VOP
mb_mode	3	Coded mode of the MB
bound	14	Position of first MB in the current
		video packet
mb_field_pred	1	Always 0 (Field DCT coding not
		supported)

Table 4-3 Motion Header

motion_req and *motion_rdy* are handshake signals. The *motion_header* signals need only be stabil, when the handshake protocol is initiated. The *MV Prediction* module registers these signals, while the differential motion vectors are being processed. This enables the *Demux* module to update the *motion_header* and the *motion_vector0-3* signals as soon as the MB header and the differential motion vectors of the next MB have been decoded. This enables the *MV Prediction* module to process the next set of differential motion vectors with a minimum delay.

4.6.2.3 Texture

The *texture_header* group of signals is decoded from the MB header and listed in Table 4-4.

Name	Width	Comments		
mb	14	MB position in current VOP		
mb_quant	5	MB quantization stepsize		
mb_mode	3	Coded mode of MB		
mb_field_dct	1	Always 0 (Field DCT coding not		
		supported)		
intra	1	1 if MB is intra-coded		
bound	14	Number of first coded MB in the		
		current video packet		
acpred_flag	1	1 if AC prediction is used		
cbp	6	Coded block pattern		
dc_diff	13	Coded DC coefficient		

Table 4-4 Texture Header

texture_req and *texture_rdy* are handshake signals. The *texture_header* signals need only be stable, when the two-way handshake protocol is initiated. The *texture_toggle* is an alternating signal that is used by the *Scan Prediction* module to detect the presence of a new texture header. The *texture_VLC* signals are used to pass the VLCs for decoding. These signals are shown in Figure 4-11.



Figure 4-11 texture_VLC Signals

If *escape* is asserted it indicates that an escape code is present and *escape_type* is used, to identify its type. The *code* signal is used to send a copy of the next 21 bits of the bitstream. When a coefficient has been decoded, the *Scan Prediction* must respond with *length* and *last*. The *Demux* module cannot continue to read the next VLC before a response from the *Scan Prediction* is received because *length* determines the position of the next VLC.

The *texture_toggle* signal is used to indicate to the Scan Prediction module that the next texture header is available. This is done to allow the DC/AC prediction to be initiated as soon as possible. Whether prediction is performed and how it is performed depends on the texture header data alone and not the coefficients.

4.6.3 Scan Prediction

This module decodes the VLCs and reconstructs the DCT coefficients of the texture MBs. The module is shown in Figure 4-12.



Figure 4-12 Scan Prediction

Once a MB has been completely reconstructed it is send to a FIFO buffer *Tbuffer*. The coefficients are clocked out one at a time via *out_data. out_mode*, *out_quant*, *out_field_dct* and *out_mb* are MB header information already described in Section 4.6.2.3.

The AC/DC prediction is initiated as soon as the *texture_header* signals are present. This is important because the DC prediction determines the scan mode to be used and consequently the MB cannot be reconstructed before the DC prediction has finished.

The *Demux* module sends a copy of the next 21 bits of the bitstream. Only 12 bits are required for regular VLC decoding, but some escape codes may require up to 21 bits. Once a VLC is decoded it is important for *Scan Prediction* to respond immediately with *length* and *last. length* is the length of the VLC and *last* indicates whether the last coefficient in a block has been decoded. The *Demux* module cannot continue to read the bitstream before this response is received. The signals used for VLC decoding is shown in Figure 4-11.

4.6.3.1 Decoding VLC

VLC decoding is implemented using the table lookup method, because this gives a fast implementation. The length of a VLC varies from 2 to 12, which requires constructing a decoding table with 4096 entries. The method is described below.

VLC CODE	LAST	RUN	LEVEL
0111	1	0	1
0000 1100 1	0	11	1
0000 0000 101	1	0	6
0011 11	1	1	1
0000 0000 100	1	0	7
0011 10	1	2	1
0011 01	0	5	1

Figure 4-13 shows a part of the VLC code table for intra coefficients

Figure 4-13 Part of the VLC Table for Intra Coefficients

Source: Reference [1]

Each entry in the decoding table has the following fields;

- LAST: A single bit boolean value to indicate end of block.
- **RUN**: A 6-bit unsigned integer indicating the number of successive zeroes preceding the coefficient.
- LEVEL: A 5-bit unsigned integer defining the coded coefficient.
- LENGTH: A 4-bit unsigned integer indicating the length of VLC (including the sign bit).

Each VLC code corresponds to one or more entries in the decoding table. The first and last entry in the table, for a specific VLC, is derived by extending the VLC with zeroes and ones respectively.

For example the VLC code "0111" from Figure 4-13 would correspond to the entries "0111 0000 0000" (1792) to "0111 1111 1111" (2047) and these entries would all contain

the field values {LAST=1, RUN=0, LEVEL=1, LENGTH=5}. The decoding table requires

 $4096 \times (1+6+5+4) = 65536$ bits = 64 Kbits

of storage.

Obviously the entries cannot overlap because a VLC cannot be a prefix of another VLC.

Decoding a VLC code is very simple once the decoding table has been constructed. First 12 bits are read from the bitstream and used to index the decoding table. The LENGTH field is used to advance the bitstream position to the start of the next VLC and the rest of the fields are used to derive the value and position of the coefficient in the current block.

MPEG-4 uses two different VLC tables for decoding coefficients, one for intra-coded MBs and another for inter-coded MBs.

Many possible EVENTs have no VLC to represent them. In order to code these statistically rare combinations an escape coding method is used. Up to 21 bits may be needed to code a coefficient in escape code.

4.6.3.2 AC/DC Prediction

Certain data for the two most recently coded rows of MBs must be stored to enable AC/DC prediction.

- 90 (6 x 15) predictors: The coefficients of the first row and column of each of the six blocks in a MB. A coefficient is a 12-bit signed integer.
- The quantization stepsize: A 5-bit unsigned integer used to scale the predictors.
- The MB type: A 3-bit unsigned integer which ensures that only intra-coded MBs are used for prediction.

AC/DC prediction is described in Section 3.1.1. For each macroblock 1088 bits of storage is required. For QCIF resolution AC/DC prediction requires

 $(2 \times 11) \times 1088 = 23936$ bits $\approx 23,3$ Kbits

of storage.

4.6.4 Tbuffer

This is a FIFO buffer that holds four complete MBs along with MB header information. A MB consists of six 8x8 blocks of 12-bit signed integers. The module is shown in Figure 4-14.



Figure 4-14 Tbuffer

Tbuffer is a two process module, which enables the buffer to be filled and emptied simultaneously. This is important because filling and emptying takes 384 clock cycles.

The storage requirement of the buffer is

$$4 \times (6 \times (8 \times 8 \times 12) + 3 + 5 + 1 + 14) = 18524$$
 bits ≈ 18 Kbits

4.6.5 Quant

This module performs inverse quantization. Both methods have been implemented. The module is shown in Figure 4-15.



Figure 4-15 Quant

A two element input buffer enables the next MB to be loaded, while the previous one is being processed. A two element output buffer enables the next MB to be processed immediately after finishing the previous one.

matrix is a group of signals from the *Demux* module and is used to load weighting matrices. These signals are shown in Figure 4-16.



Figure 4-16 Matrix signals

req and *rdy* are handshake signals. *intra* indicates whether an intra or an inter matrix is being loaded and *coeff* transmits the matrix elements. The process doing the inverse quantization also handles loading the matrices, since the quant module is idle, when a new matrix is being loaded and hence no unnecessary latency is introduced.

4.6.6 Idct

This module performs the Inverse Discrete Cosine Transformation on each of the six blocks in a macroblock. The module is shown in Figure 4-17.



Figure 4-17 Idct

The internal structure is similar to the Quant module using both input and output buffers.

The 2D IDCT from Section 3.1.3 is repeated below for readability

$$f(x, y) = \frac{1}{4} \sum_{u=0}^{7} \sum_{v=0}^{7} C(u)C(v)F(u, v)\cos\frac{(2x+1)u\pi}{16}\cos\frac{(2y+1)v\pi}{16}$$
$$C(u), C(v) = \begin{cases} \frac{1}{\sqrt{2}} & \text{for } u, v = 0\\ 1 & \text{otherwise} \end{cases}$$

f(x,y) are the original 8 bit pixel values and F(u,v) are the DCT coefficients.

The 2D IDCT can be written as

$$f(x,y) = \sum_{\nu=0}^{7} \frac{1}{2} C(\nu) \left(\sum_{u=0}^{7} \frac{1}{2} C(u) F(u,\nu) \cos \frac{(2y+1)u\pi}{16} \right) \cos \frac{(2x+1)\nu\pi}{16}$$

The 2D IDCT can be decomposed into two 1D IDCT using row and column decomposition. Basically the 2D-IDCT can be decomposed into a 1-IDCT on the rows followed by a 1D-IDCT on the columns. Both phases require eight 1D-IDCTs to be calculated. These can be calculated in parallel and require 64 iterations. The decomposed version of the IDCT requires only 128 iterations as opposed to the 4096 needed for the 2D-IDCT. The implemented algorithm was proposed by Arai, Agiu and Nakajima [5].

The 1D-IDCT is a four-stage process. Table 4-5 shows the number of operations for each stage.

	Shifts	Additions	Multiplications
Stage 1	2	7	6
Stage 2	0	9	3
Stage 3	2	7	2
Stage 4	8	8	0
Total	12	31	11

Table 4-5 Operations used in 1D-IDCT

Pipelining the IDCT loop can improve overall latency significantly.

4.6.7 MV Prediction

This module receives the differential motion vectors decoded from the bitstream and adds the prediction to calculate the final motion vectors. The reconstructed motion vectors are send to a FIFO buffer *Mbuffer*. The module is shown in Figure 4-18.



Figure 4-18 MV Prediction

in_req, *in_rdy*, *out_req* and *out_rdy* are handshake signals. *mvd0-3* are the differential motion vectors extracted from the bitstream and *mv0-3* are the final motion vectors. The rest of the signals are MB header signals described in Section 4.6.2.2.

Four motion vectors for the two most recently coded rows of macroblocks must be stored to enable vector prediction. For each macroblock 104 bits of storage is required. For QCIF resolution vector prediction requires

$$(2 \times 11) \times 104$$
 bits = 2288 bits \approx 2,2 Kbits

of storage.

4.6.8 Mbuffer

This is a FIFO buffer that holds four elements. Each element contains motion vectors for a MB along with MB header information. The module is shown in Figure 4-19.


Figure 4-19 Mbuffer

The module is very similar to *Tbuffer*. The storage requirement of *Mbuffer* is $4 \times ((4 \times 2 \times 13) + 10 + 10 + 3) = 508$ bits

which is significantly less than Tbuffer.

4.6.9 Interpolate

This module interpolates the reference VOP using the decoded motion vectors. The module is shown in Figure 4-20.



Figure 4-20 Interpolate

The coefficients of the interpolated MB are clocked out to the *Reconstruct* module via the *out_data* port. The rest of the ports connected the Reconstruct module are the usual handshake signals and MB header signals. Input and output buffers are used to minimize latency.

The reference VOP is accessed through the *Memory Controller*. *req* and *rdy* are handshake signals. Handshake signals are used so that the reference VOP is not accessed, while it is being updated. *ref_x*, *ref_y* and *yuv* are used to request a 9x9 block from the reference VOP. *yuv* is used to indicate, which color component to access; 0=y, 1=u, 2=v. *ref_x* and *ref_y* are the coordinates of the upper left corner of the block referenced. The *Memory Controller* accesses the memory through a 32-bit data bus, which means that byte access is not possible. Instead the *Memory Controller* returns a block, which is guaranteed to contain the referenced block. Basically a sequence of 27 four-byte words representing a 12x9 block is returned from the reference VOP. The 9x9 block. The two least

significant bits of ref_x determines the offset of the 9x9 block within the returned block. This is shown in Figure 4-21.



Figure 4-21 a) Sequence of 27 Words b) 12x9 Block

The upper left corner of the 9x9 block for different offset values is shown. The grey box is the 9x9 block for offset equal to 2. Only the coefficients within the 9x9 block are stored in registers for further processing, the rest of the coefficients are ignored. When the horizontal component of the motion vector is full-pel the right-most column of the 9x9 block is not used, and when the vertical component of the motion vector is full-pel the motion vector is full-pel the bottom row of the 9x9 block is not used.

Although a 32-bit data bus complicates access to the reference VOP during interpolation, data transfers can be done using significantly less clock cycles. Once the memory controller has set up the transfer, it takes only 27 clock cycles to transfer a 12x9 block. An 8-bit data bus would enable transfer of only the necessary 9x9 block, however this would require 81 clock cycles. Additionally updating the reference VOP requires four times less clock cycles, when using a 32-bit data bus.

The *Interpolate* module stores a complete 9x9 block of coefficients in registers before performing the interpolation. Alternatively the coefficients could be read while performing the interpolation. But this would lead to inefficient synthesis of the interpolation loop. This is explained below.

The interpolation loop is a prime candidate for pipelining. A code fragment of the interpolation loop, when both components of the motion vector are half-pel, is shown below.

```
for(j=0; j<8; j++){
    for(i=0; i<8; i++){
        block[j*9+i] =
            ((block[j*9+i] + block[j*9+i+1] + block[j*9+i+9] + block[j*9+i+9+1] + 2 - rounding) >> 2);
    }
}
```

Four reference samples are needed to calculate a single interpolated sample, when both components of the motion vector are half-pel. If only one component is full-pel two reference samples are needed. If both components are full-pel no interpolation is performed.

The loop has no loop carry dependencies, no memory or I/O accesses and the exit condition is evaluated in the first clock cycle of the loop. This loop can be pipelined with an initiation interval of 1, which means that loop iterations finish one clock cycle apart resulting in optimal throughput.

Reading the samples from the reference VOP, while performing interpolation was considered, but dismissed due to the two-way handshake protocol used for communication with the *Memory Controller*. Eliminating the two-way handshake and rewriting the *Memory Controller* to operate with fixed latency would still imply a latency through the *Memory Controller* of 3, which would not produce the optimal throughput described above.

4.6.10 Reconstruct

The *Reconstruct* module receives MBs resulting from texture and motion decoding and stores them correctly in the frame buffer. The frame buffer is accessed through the *Memory Controller*. The module is shown in Figure 4-22.



Figure 4-22 Reconstruct

The motion and texture part of the module do basically the same. *Fill Texture Buffer* fills the texture input buffer, and *Fill Motion Buffer* fills the motion buffer. *Entry* empties the motion and texture buffer and reconstructs the MBs, which are then send to the *Memory Controller*.

The *tail* signal is used to indicate that a new MB has been inserted in the buffer. The *update* signal is used to indicate that a MB has been read from the buffer. *Buffer Controller* uses *tail* and *update* to derive the *head* and *is_empty* signals. *head* is the position of the next MB in the buffer and *is_empty* indicates the status of the buffer.

Basically, the texture buffer and the motion buffer are independently filled, and the *entry* process empties the buffers according to the scheme described next.

Intra-coded texture MBs and not-coded motion MBs are immediately send to the *Memory Controller* because they contain the full data for the MB. When an inter-coded texture MB is detected the corresponding motion MB is added before sending it to the *Memory Controller*. This motion MB may be preceded by not-coded MBs in the motion buffer, which must also be send to the *Memory Controller*. When an inter-coded motion MB is added before sending it to the *Memory Controller*. This texture MB may be preceded by intra-coded MBs in the texture buffer, which must also be send to the *Memory Controller*. When an inter-coded motion MB is added before sending it to the *Memory Controller*. This texture MB may be preceded by intra-coded MBs in the texture buffer, which must also be send to the *Memory Controller*. An example of how a sequence of macroblocks is stored in the buffers is shown in Figure 4-23.



Figure 4-23 a) The decoded sequence. b) The texture sequence. c) The motion sequence.

The above scheme ensures that all macroblocks are stored in the frame buffer and that the texture and motion parts of inter-coded macroblocks are added correctly.

4.6.11 Memory Controller

This module controls access to the reference VOP and the frame buffer. The module is shown in Figure 4-24.



Figure 4-24 Memory Controller

The *Reference VOP* and the *Frame Buffer* are modeled as simple dual port SRAM memory blocks. The *Frame Buffer* holds 4 VOPs. The *Port* signals are shown in Figure 4-25.



Figure 4-25 Memory Block Interface

Requests from the *Interpolate* module are handled by *Read Ref. Read Ref* uses *x*, *y*, and *yuv* to calculate the physical address of the requested block in the reference VOP and sends the block back to the *Interpolate* module.

Send Frame reads the content of the oldest frame in the *Frame Buffer* and sends it to the *Display* module. The interface signals to the *Display* module will be described in Section 4.6.12.

Build Frame stores the reconstructed MBs in the frame buffer. If the frame buffer is full the *Reconstruct* module is forced to wait. Once *Build Frame* has stored the last MB in a VOP, it initiates a handshake communication with the *Demux* module. This is done to signal that the decoder is idle, i.e. the decoder is no longer processing any MBs. This will allow the *Demux* module to safely update the control signals and initiate an update of the reference VOP. The *update_ref* group of signals, which is used to communicate with the *Demux* module is shown in Figure 4-26.



Figure 4-26 Update_ref Signals

The *update_ref_req* and *update_ref_ack* are handshake signals. *update _ref* is a boolean signal, which indicates whether the next VOP is an P-VOP, i.e. whether to update the reference VOP. When an update of the reference VOP is initiated, *Build Frame* handles the data transfer between the frame buffer and the reference VOP as well as the padding process. This allows the *Demux* module to continue reading the bitstream. While the reference VOP is being updated the *Read Ref* process is blocked.

4.6.12 Display

This module converts the YUV frame stored in the frame buffer to a sequence of RGB values. The RGB values are output line by line. The module is shown in Figure 4-27.



Figure 4-27 Display

Get YUV fills the input buffer with YUV samples from the frame buffer. Each element in the input buffer contains 16 luminance samples and 16 chrominance samples. *Convert* then converts the YUV samples to RGB values and writes them to the output buffer. The buffer structures are shown in Figure 4-28.



Figure 4-28 a) Input buffer element b) Output buffer element

Luminance samples are read line by line and 16 samples at a time. Chrominance samples are also read line by line, but only 8 samples at a time. Each chrominance line is read

twice so that the chrominance samples correspond to the luminance samples. The *Memory Controller* handles access to the frame buffer and ensures that the proper samples are transferred to the *Display* module.

Send RGB sends the RGB samples to the Output Generator module one pixel at a time.

4.6.12.1 YUV to RGB

The conversion formula is repeated here for readability.

b	=	1.164 (y – 16)			+	2.018 (<i>u</i> – 128)
g	=	1.164 (y – 16)	_	0.813 (v – 128)	_	0.391 (<i>u</i> – 128)
r	=	1.164 (y – 16)	+	1.596 (v – 128)		

The conversion is done using fixed point arithmetic and predefined lookup tables. 13 bits are used for the fractional part and the following constants are predefined.

```
#define SCALEBITS_OUT 13#define FIX_OUT(x)((uint16_t) ((x) * (1L<<SCALEBITS_OUT) + 0.5))</td>#define RGB_Y_OUTFIX_OUT(1.164)#define B_U_OUTFIX_OUT(2.018)#define G_U_OUTFIX_OUT(0.391)#define G_V_OUTFIX_OUT(0.813)#define R_V_OUTFIX_OUT(1.596)
```

Five lookup tables are then predefined.

for (i = 0; i < 256; i++) {	
RGB_Y_tab[i]= RGB_Y_OUT	* (i - Y_ADD_OUT);
$B_U_tab[i] = B_U_OUT$	* (i - U_ADD_OUT);
G U tab[i] = G U OUT	* (i - U_ADD_OUT);
G_V_tab[i] = G_V_OUT	* (i - V_ADD_OUT);
R V tab[i] = R V OUT	* (i - V ADD OUT);
}	//

The conversion of each pixel requires 5 lookups, 4 additions and 3 shifts.

```
b_u = B_U_tab[u];
g_uv = G_U_tab[u] + G_V_tab[v];
r_v = R_V_tab[v];
rgb_y = RGB_Y_tab[y];
b = (rgb_y + b_u) >> SCALEBITS_OUT;
g = (rgb_y - g_uv) >> SCALEBITS_OUT;
r = (rgb_y + r_v) >> SCALEBITS_OUT;
```

The values are clamped to keep them within the range [0; 255].

4.6.13 Output Generator

This module outputs the RGB values to a monitor. The module is shown in Figure 4-29.



Figure 4-29 Output Generator

RGB values are read into a buffer and sub sequentially written to a monitor. Each buffer element contains the RGB samples of a single pixel. *Sync Signal* provides the signal that controls the output rate. *valid* is asserted when a new output is written. *new_line* is asserted to indicate the start of a new line. *new_frame* is asserted to indicate the end of the last line.

This module has been implemented with testing in mind. The output signals need to be altered for frames to be shown on a display.

The *monitor* is part of the testbench and is described in Chapter 5.

5 Testing

The decoder has been tested for correctness at the behavioral level. The testbench shown in Figure 5-1 was used.



Figure 5-1 Testbench

The stimulus generator is the memory holding the MPEG-4 bitstream. The content of the memory is read from a file holding the bitstream. The format of this file is described in Section 5.1. The monitor stores the output of the decoder as a sequence of uncompressed BMP image files. A full description of the BMP file format is included in Appendix D. Appendix F includes a description of how to compile and simulate the model.

5.1 Bitstream storage

The MPEG-4 bitstream used for testing was stored in a MP4U container file. The file format is used by XviD developers for testing purposes and has the file extension "*.mp4u*". The initial 4 bytes of the file contains the string "MP4U" and the rest of the file contains chunks of data. Each chunk of data contains the coded bitstream for a single VOP. Data chunks are preceded by a 4 byte field, which indicates the size of the data chunk. The "MP4U" string is used to identify a valid MPEG-4 bitstream container and the size field enables users to import a fraction of the bitstream at a time. A MP4U

container file is used by the stimulus generator to emulate the content of the bitstream memory.

5.2 Bitstream creation

A program "avi2xvid" capable of extracting the MPEG-4 bitstream from an AVI file was written. The program extracts the MPEG-4 bitstream from any AVI file coded with the XviD CODEC and stores it in a MP4U container file. Additionally the program was linked to the original unmodified XviD CODEC so that the bitstream could be decoded. The XviD CODEC support several colorspace formats for the decoded frames. The decoded frames were stored as a series of BMP image files in the 24-bit RGB colorspace. These BMP files were later compared with the decoder output to test the correctness of the decoder. It is important to note that the RGB values stored in the BMP files were the direct output of the original unmodified XviD CODEC.

Encoding software VirtualDub 1.5.1 was used to encode AVI files with the XviD CODEC. The encoder settings are listed in Appendix B.

A brief description of the AVI file format is included in Appendix E.

5.3 Test results

VirtualDub was used to generate a test video containing 10 frames coded with a constant bit rate of 64 Kbps using H.263 quantization and a resolution of 144x176. The first frame was intra coded and the following 9 frames inter coded. The encoder settings used to encode the test video is listed in Appendix B. "avi2xvid" was used to extract the bitstream and the decoded frames from the test video. The generated MP4U container file was then used in the testbench.

Finally the decoder was used to decode the bitstream and the output compared with the output of the original XviD CODEC. The comparison was done using simple file comparison. The output files were identical to the reference files generated by the

unmodified XviD CODEC. The test successfully verified the correctness of the implemented decoder.

5.4 Bandwidth analysis

Real-time decoding of QCIF resolution video at 15 fps requires decoding of 9x11 MBs 15 times a second. Each MB requires 6x8x8 coefficients to be processed. The bandwidth requirements for the image processing modules; *Quant, Idct* and *interpolate* are

 $9 \times 11 \times 15 \times (6 \times 8 \times 8) = 570240$ coeff./s

Assuming a clock cycle of 50 ns (20MHz) means that each coefficient in a MB must be processed in

 $20 \times 10^6 \text{ s}^{-1} / 570240 \text{ s}^{-1} = 35,0730 \approx 35 \text{ clock cycles}$

Each 8x8 block must be processed within

 $35 \times 8 \times 8 = 2240$ clock cycles

5.5 Synthesis

The *Tbuffer* module was scheduled, and a structural RTL netlist, ready for compilation to gates, was generated. Additionally the scheduled design was written to a SystemC format file optimized simulation speed. This file contained a cycle-accurate RTL-netlist and was used for cycle-accurate simulation of the *Tbuffer* module in the testbench. The simulation verified the correctness and synthesizability of the *Tbuffer* module. This was important because the internal buffer structure and the interface protocol of the Tbuffer module was used in most of the other modules.

The *Quant* module was also synthesized to a structural RTL netlist. A scheduling report was generated and an analysis of the report showed that 12 clock cycles is used for the quantization of a single coefficient, without pipelining. This is well within the 35 clock cycles available.

The *Idct* module was not synthesized. An implementation of the IDCT on a Xilinx Virtex FPGA is describes in [16]. This implementation uses 16 1D-IDCT hardware modules. Each 1D-IDCT module has latency of 13 clock cycles at a clock frequency of approximately 53 MHz. An implementation of the 8x8 IDCT using only a single of these modules would result in a block latency of

 $2 \times (8 \times 13) = 208$ clock cycles

This is well within the 2240 clock cycles available.

6 Conclusion

An MPEG-4 Simple@L0 compliant decoder has been successfully modeled using SystemC. The model has been tested for correctness at the behavioral level. The test was performed by comparing the output of the model with the output of the XviD CODEC.

The rewriting process of converting the XviD CODEC to SystemC proved more time consuming than originally anticipated. The structural differences of a software codec and a hardware decoder required a lot of rearranging. XviD handles intra frames and inter frames using separate functions. These functions had to be merged and integrated into modules capable of processing both intra and inter frames. Data size and access patterns also required careful analysis. XviD declares storage for data structures such as the bitstream, VOP header, MB header, and the reference VOP at the top hierarchical level, and then passes pointers to access these. In hardware modeling, modules only have access to their own data. Data defined in other modules must be transferred before it can be accessed, which introduces additional latency. Declaration of data storage was done with the aim of minimizing data transfers. Replacing the native C++ data types with bit width accurate data types required in some cases very careful analysis of their range.

The most time consuming parts of the rewriting process were elimination of pointers, declaration of data storage and integration of intra and inter functions.

Behavioral modeling requires use of two-way handshake protocols because designs are not cycle accurate. This made it more difficult to model interfaces because clock accurate behavior could not be determined before synthesis. Modeling the internal behavior of the modules was on the other hand simple once the intra and intra parts had been rearranged into modules.

The decoder was completely modeled and the correctness verified before an attempt was made to synthesize it. It became apparent that some of the code, although not violating any of the design rules in the Synopsys documentation, was not synthesizable. Sometimes the synthesis tools would display fatal error messages with no indication about the source of the problem. Other error messages referred to lines in the code not actually causing the problem. This made it difficult to identify the source of the problem. Often the only option was to comment out code until the problem was identified. Although the errors often turned out to be caused by non-synthesizable code, more time was spent identifying the problems, rather than fixing them. A better approach would have been to synthesize each module as they were written. This would have reduced the repeated use of the same non-synthesizable code in other modules.

One of the benefits of behavioral synthesis is the ability to explore the design space without having to change the source code significantly. Experimenting with speed, power and area tradeoffs, for modules critical to overall latency would have been a very interesting part of the project. However most of the time was spend on making the modules synthesizable. Using the pipelining tool to improve throughput would also have been interesting. Image processing is highly repetitive and involves many iterations, which makes it ideal for pipelining.

The *Tbuffer* module was successfully synthesized to a structural RTL netlist and simulated at cycle accurate level. This module was chosen as the first module to be synthesized, because it is simple and because it uses the input/output buffer structure used in several other modules. Additionally, the two-way handshake protocol, which is used to transfer MBs between modules, was used by this module.

Behavioral synthesis schedules operations into clock cycles. This higher level of abstraction offered by behavioral modeling is definitely useful for modeling algorithms. However behavioral modeling of interfaces and communication protocols does not offer the same control as RTL modeling. A process cannot release a shared resource, such as a data bus, using high impedance signals. In fact only one process is allowed to drive a signal in behavioral modeling. Combinatorial modules such as multiplexers are not

allowed, which means that all modules have an inherent latency of at least one clock cycle.

Future work on the model may include implementation of an input buffer scheme for importing the bitstream. Implementation of a DRAM interface for the reference VOP and the frame buffer memory blocks. Implement a module capable of displaying the decoded frames on a display. Full synthesis of the decoder to FPGA using the pipeline tool.

It has been very interesting to learn about MPEG-4 and the compression tools included in the standard. Although not relevant for this project the object-based representation of audiovisual content adopted by MPEG-4 is also an area that seems very exciting. Greater understanding and appreciation of the issues involved in block-based DCT compression methods and valuable experience in behavioral modeling was achieved. Dealing with coded MPEG-4 data at bit level gave valuable insights into how bitstreams are constructed.

End of report

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7 Appendix A

MPEG-4 Background

Earlier standards such as MPEG-1 and MPEG-2 addressed the issue of compression by representing video as a sequence of rectangular frames. The primary goal of these standards was efficient compression. The developers of the MPEG-4 standards met for the first time in September 1993 to identify the requirements of a very low bit-rate coding solution. It was soon realized that results such as those produced by the LBC (Low Bit-rate Coding) (later known as the ITU-T H.263) group where close to the optimal using the traditional block-based, hybrid, DCT/motion-compensation video decoding method.

Objectives

At the time it was believed that significant improvements over the LBC standard using conventional coding techniques that would justify another standard could not be achieved. The group felt the need to broaden the objectives of the MPEG-4 standards.

They assessed current and future trends in the world of multimedia and specifically addressed the issue by observing how audiovisual content was produced, delivered and consumed.

Traditionally audiovisual content where mostly natural 2D content produced with cameras and microphones and edited at the production stage. The audiovisual content would most often be transmitted over a few homogenous networks (ISDN, LAN, Satellite) using their own representation scheme. Consumption would be passive and the content pre-composed.

The group anticipated future audiovisual content to include both natural and computergenerated synthetic elements and produced such that coding and composition is done separately for individual objects. Delivery would happen across heterogeneous networks with varying bandwidth capabilities using a common representation scheme. Some degree of composition would happen at the consumption stage allowing information to be consumed interactively.

These observations resulted in a coding standard which would support content-based communication, access and manipulation of digital audiovisual data. The aim was to achieve both high interactivity and better compression ratio. Where as previous standards viewed video as sequence of rectangular frames the MPEG-4 standards adopted another approach. An object-based representation of audiovisual content was adopted.

This approach implied building an audiovisual scene from individual object related only in time and space. This would enable coding of different types of objects using the most efficient representation of that particular type of object. Object types may include rectangular objects, arbitrarily shaped objects, music, speech, synthetic objects, text, still pictures etc. MPEG-4 standards include special synthetic objects such as 2D and 3D objects and face and body animation objects.

The object-based approach allow coding of multiple concurrent data streams, mixing of natural and synthetic objects and content based scalability.

For example a user can edit or change certain characteristics of a single object in the scene by manipulating the bit stream directly without ever having to transcode the sequence. Scalability implies the ability to select the decode quality or to prioritize among object in a scene, useful for delivery over heterogeneous network environments.

Early video coding standards where developed with efficient compression in mind and have been successfully adopted for commercial use including digital broadcasting, VCD, DVD and digital television to mention a few. The MPEG-4 standard introduces a new way of representing audiovisual data using an object-based approach, which is similar to the way humans view the visual world. This approach allows content based access to data and a level of interactivity not previously seen in conventional coding schemes.

XviD

Originally Microsoft developed a MPEG-4 standard compliant codec called Windows Media Video V3 (WMV). Microsoft's WMV encoder did not support saving MPEG-4 streams to the AVI file format, forcing users to use Microsoft's own file format Advanced Streaming Format (ASF). The ASF file format also restricted users to use Microsoft's Windows Media Audio (WMA) encoder codec to save audio streams. These restrictions lead to the development of Divx;) also known as DivX v3.11 alfa, which was a hacked illegal version of the original WMV codec. DIVX;) supported saving to AVI files as well as mp3 audio streams and quickly became very popular.

Its popularity lead the people behind the hacked version to start up new a company DivXNetworks Inc in May 2000 and development of a legal version of the codec commenced. This coded would be coded from scratch and began as an open source project called Project Mayo or OpenDivX. The company eventually started developing a closed source version of the codec and released it in August 2001 under the name DivX 4.0.

Open source developers from the original Project Mayo project wished to continue an open source version and started the XviD (DivX spelled backwards) project.

When this project started the latest version was version 0.91 and this version is used for this project. This version implements the Simple Profile. Current versions of XviD implement the Advanced Simple Profile.

8 Appendix B

8.1 Encoder settings

VirtualDub 1.5.1 was used as the encoding software in this project. Table 8-1 lists the encoder settings. The 2^{nd} column indicates the available settings provided by the encoding software. The 3^{rd} column indicates the restrictions to these settings by the Simple@L0. The final column indicates the settings used to encode the test video used for testing. Empty fields indicate no restrictions.

Setting	Range	Simple@L0	Test Video
Frame rate (fps)	≥1	Max. 15	15
Encoding mode	1 Pass – CBR		1 Pass - CBR
	1 Pass – quality		
	1 Pass – quantizer		
	$2 \text{ Pass} - 1^{\text{st}} \text{ pass}$		
	$2 \text{ Pass} - 2^{\text{nd}} \text{ pass ext.}$		
	$2 \text{ Pass} - 2^{\text{nd}} \text{ pass int.}$		
Bitrate (Kbits/s)	≤10000	64	64
Motion search	0 – None		6 – ultra high
precision	1 - very low		
	2 - low		
	3 – medium		
	4 – high		
	5 – very high		
	6 – ultra high		
Quantization type	H.263	H.263	H.263
	MPEG		
	MPEG-Custom		
Max. I-frame interval	≥0	≥0	11
Min. I-frame interval	≥ 0	≥ 0	11
Luminance masking	Disable		Disable
	Enable		
Interlacing	Disable	Disable	Disable
	Enable		
Min. I-frame quantizer	1-31	1-31	1
Max. I-frame quantizer	1-31	1-31	31
Min. P-frame quantizer	1-31	1-31	1
Max. P-frame quantizer	1-31	1-31	31
Custom intra matrix	Disable	Disable	Disable
	Enable		
Custom inter matrix	Disable	Disable	Disable

	Enable		
Max. bitrate (Kbits/s)	≥Bitrate	≤64	
Decomp. format	16-bit (HiColor)	24-bit (TrueColor)	
	24-bit (TrueColor)		
Output format	16-bit (HiColor)	24-bit (TrueColor)	
	24-bit (TrueColor)		
	32-bit (TrueColor)		

Table 8-1 Encoder settings

9 Appendix C

VOP Header

Table 9-1 lists the data elements included in the VOP header.

Name	width	Comments
vop_start_code	32	000001B6 (Hexadecimal)
vop_coding_type	2	VOP coding method. Defined in Table 9-2
modulo_time_base	-	Consists of a number of consecutive 1 followed by a 0.
		The number of 1s indicate the number of seconds elapsed
		since the last synchronization point
vop_time_increment	1-16	The absolute time increment from the synchronization
		point marked by modulo_time_base measured in clock
		ticks
vop_coded	1	When set to 0 no subsequent data exists for the VOP.and
		the VOP is copied from the reference VOP.
vop_rounding_type 1 U		Used for proper rounding during interpolation
intra_dc_vlc_thr 3		A 3-bit code which is used to switch between two VLCs
		for coding of intra DC coefficients. Defined in Table 9-3
quant_type	1	Determines which of the two quantization methods to use.
vop_quant	5	An unsigned integer which specifies the absolute
		quantizer stepsize to be used for dequantizing the MB
		until updated by dquant
vop_fcode_forward	3	An unsigned integer used for decoding motion vectors

Table 9-1 - VOP header

vop_coding_type	coding method
00	intra-coded (I)
01	predictive-coded (P)
10	bidirectionallly-predictive-coded (B)
11	sprite (S)

Table 9-2 Meaning of vop_coding_type

Source: Reference [1]

index	meaning of intra_dc_vlc	code
0	use Intra DC VLC for entire VOP	000
1	switch to Intra AC VLC at running Qp≥13	001
2	switch to Intra AC VLC at running Qp≥15	010
3	switch to Intra AC VLC at running Qp≥17	011

4	switch to Intra AC VLC at running Qp≥19	100
5	switch to Intra AC VLC at running Qp≥21	101
6	switch to Intra AC VLC at running Qp≥23	110
7	use Intra AC VLC for entire VOP	111

Table 9-3 Meaning of intra_dc_vlc_thr

Source: Reference [1]

running Qp is the quantization stepsize of the previous MB.

MB Header

Table 9-4 lists the data elements included in the MB header.

Name	width	Comments			
not_coded	1	When set to 1 no more data exists for the current macroblock			
		and the macroblock is copied from the reference VOP			
mcbpc	1-9	A VLC that is used to derive the macroblock type and the			
		coded block pattern for chrominance			
acpred_flag	1	When set to 1 AC prediction is used			
cbpy	1-6	A VLC that is used to derive the coded block pattern for			
		luminance (Y)			
dquant	2	A 2-bit code specifies changes to vop_quant. Found only in			
		INTRA_Q and INTER_Q MBs. The changes are valid for all			
		subsequent MBs. Defined in Table 9-6			
dc_scaler*	6	The dc_scaler may be different for luminance and			
	chrominance components.				
quant_stepsize*	5	Quantization stepsize.			

Table 9-4 MB header

Source: Reference [1]

**quant_stepsize* and *dc_scaler* are derived from the VOP header and the MB header and are not coded in the bitstream.

In addition to the data elements listed in Table 9-4 each MB is associated with a quantization stepsize called *quant_stepsize*. *quant_stepsize* is equivalent to *vop_quant* after the changes by *dquant* have been made.

dc_scaler is derived from quant_stepsize. This is shown in Table 9-5.

Component	dc_scal	er for quant_stepsiz	æ	
	1 to 4	5 to 8	9 to 24	≥25
Luminance	8	2* quant_stepsize	quant_stepsize+8	2* quant_stepsize-16
Chrominance	8	(quant_stepsize+13)/2	quant_stepsize-6

Table 9-5 dc_scaler expressed in terms of the quant_stepsize

Source: Reference [1]

dquant code	value
00	-1
01	-2
10	1
11	2

Table 9-6 dquant codes

Source: Reference [1]

10 Appendix D

10.1 BMP file format

The BMP file format is described next. Only those parts relevant for this project are described. The Device Independent Bitmap is the format which is used to store bitmaps in BMP files. A BMP file contains the following structures

- Bitmap file header
- Bitmap info header
- RGB values

The bitmap file header contains information about the BMP file. The bitmap info header contains information about the bitmap data. The RGB values are the image data. The structure of the bitmap file header is shown in Table 10-1. All sizes are in byte units.

Name	Size	Default	Comments
bfType	2	19778	Always "BM"
bfSize	4	-	Size of the file in bytes
bfReserved1	2	0	Always 0
bfReserved2	2	0	Always 0
bfOffBits	4	54	Offset from beginning of file to image data

Table 10-1 Bitmap file header

The structure of the bitmap info header is shown in Table 10-2.

Name	Size	Default	Comments
biSize	4	40	Size of bitmap info header
biWidth	4	-	Width of image in pixels
biHeight	4	-	Height of image in pixels
biPlanes	2	1	Always 1
biBitCount	2	24	Number of bits per pixel
biCompression	4	0	Always 0 (No compression)
biSizeImage	4	0	Always 0 (No compression)
biXPelsPerMeter	4	0	Horizontal pixels per meter on the target device
			(usually set to zero)
biYPelsPerMeter	4	0	Vertical pixels per meter on the target device
			(usually set to zero)
biClrUsed	4	0	Number of colors used in the bitmap. Automatically
			calculated when set to 0

biClrImportant	4	0	Number of colors that are important for the bitmap. Set to 0 when all colors are important

Table 10-2 Bitmap info header

The image data are stored a row at a time starting from the bottom row. The start of each row is aligned to the next 4-byte position. Each pixel in a row is represented as a sequence of B, G, and R values in that order.

11 Appendix E

11.1 AVI File format

The AVI file is a RIFF file containing mandatory RIFF chunks. The top-level chunk is the AVI RIFF chunk with a FOURCC of AVI. The AVI RIFF chunk includes two mandatory list chunks with FOURCC *hdrl* and *movi* and an optional index chunk with FOURCC *idx1*. The *hdrl* chunk includes header information about the format of the included data. The *movi* chunk contains the actual data, and *idx1* is an optional index of the data. For more information on the RIFF file format readers are referred to [22].
12 Appendix F

12.1 File structure

The source files are listed in Table 12-1.

Module	Location	Files	Comments
-	mpeg4\	Makefile	builds the
			decoder
-	mpeg4\	constants.h	constants
-	mpeg4\	global.h	data types
-	mpeg4\	main.cpp	testbench (top-
			level)
getBits	mpeg4\getbits\	getbits.h	
		getbits.cpp	
Demux	mpeg4\demux\	demux.h	
		demux.cpp	
Scan	mpeg4\texture\	scan_prediction.h	
Prediction		scan_prediction.cpp	
Tbuffer	mpeg4\texture\	tbuffer.h	
		tbuffer.cpp	
Quant	mpeg4\texture\	quant.h	
		quant.cpp	
Idct	mpeg4\texture\	idct.h	
		idct.cpp	
MV Prediction	mpeg4\motion\	mv_prediction.h	
		mv_prediction.cpp	
Mbuffer	mpeg4\motion\	mbuffer.h	
		mbuffer.cpp	
Interpolate	mpeg4\interpolate\	interpolate.h	
		interpolate.cpp	
Reconstruct	mpeg4\reconstruct\	reconstruct.h	
		reconstruct.cpp	
Memory	mpeg4\memory_controller\	memory_controller.h	
Controller		memory_controller.cpp	
Display	mpeg4\display\	display.h	
		display.cpp	
Output	mpeg4\display	output_generator.h	
Generator		output_generator.cpp	
Memory	mpeg4\memory\	memory.h	bitstream
		memory.cpp	
Monitor	mpeg4\monitor\	monitor.h	writes to bmp
		monitor.cpp	
Frame Buffer	mpeg4\ram\	cur_ram.h	
		cur_ram.cpp	

Reference	mpeg4\ram\	ref_ram.h	
VOP		ref_ram.cpp	

Table 12-1 Source files

Other files are listed in Table 12-2.

File	Location	Comments
out.mp4u	avi2xvid\	MP4U container with the coded
		bitstream for the test video
test_video	avi2xvid\	The AVI test video
Makefile	avi2xvid\	Makefile for avi2xvid
avi2xvid.exe	avi2xvid	Program
lixvidcore.lib	avi2xvid\	XviD static link library
-	xvidcore\src\	XviD CODEC source filer
Makefile	xvidcore\build\	Make file for XviD CODEC

Table 12-2 Other files

12.2 Simulating the model

Build the XviD CODEC by running

>xvidcore\build\generic\configure

>xvidcore\build\generic\make

>xvidcore\build\generic\make install

This will build and install the static link library xvidcore.lib to av2xvid\.

Then build avi2xvid by running >avi2xvid\make

Then run

>avi2xvid test_video.avi

This will create the out.mp4u and the BMP image files in the avi2xvid\frames\ directory. Copy the out.mp4u to mpeg4\ and run >mpeg4\make

This will build the MPEG-4 decoder simulation model. Run >mpeg\decoder 3000000

The parameter is the maximum number of cycles used in the simulation. The model will now store the decoded frames in mpeg4\frames\ directory. These can be compared with the frames in avi2xvid\frames\.